

# Global Optimal Realizations of Finite Precision Digital Controllers

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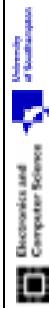
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## Problem Definition

- Plant:  $P(z) \sim (\mathbf{A}_P, \mathbf{B}_P, \mathbf{C}_P); \mathbf{A}_P \in \mathcal{R}^{m \times m}, \mathbf{B}_P \in \mathcal{R}^{m \times l}, \mathbf{C}_P \in \mathcal{R}^{q \times m}$ .
- Controller:  $C(z) \sim (\mathbf{A}_C, \mathbf{B}_C, \mathbf{C}_C, \mathbf{D}_C); \mathbf{A}_C \in \mathcal{R}^{n \times n}, \mathbf{B}_C \in \mathcal{R}^{n \times q}, \mathbf{C}_C \in \mathcal{R}^{l \times n}, \mathbf{D}_C \in \mathcal{R}^{l \times q}$ .

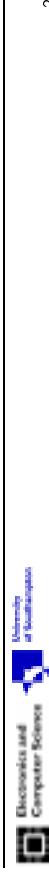
Denote an initially designed controller realization as  $\mathbf{X}_0$  and a generic realization  $\mathbf{X}$ . Let  $\overline{\mathbf{A}}(\mathbf{X})$  be the closed-loop transition matrix with  $\mathbf{X}$ .

- Controller realization set

$$\mathcal{S}_C \triangleq \{\mathbf{X} : \mathbf{A}_C = \mathbf{T}^{-1}\mathbf{A}_C^0\mathbf{T}, \mathbf{B}_C = \mathbf{T}^{-1}\mathbf{B}_C^0, \mathbf{C}_C = \mathbf{C}_C^0\mathbf{T}, \mathbf{D}_C = \mathbf{D}_C^0\}$$

where  $\mathbf{T} \in \mathcal{R}^{n \times n}$  is an arbitrary non-singular matrix

- All  $\mathbf{X} \in \mathcal{S}_C$  are equivalent in infinite precision implementation: an identical set of closed-loop eigenvalues  $\lambda_i(\overline{\mathbf{A}}(\mathbf{X}))$ ,  $1 \leq i \leq m+n$ , which are all within the unit disk.



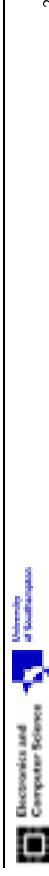
## Closed-Loop Stability Robustness Measure

- In FWL fixed-point implementation,  $\mathbf{X} \rightarrow \mathbf{X} + \Delta\mathbf{X}$  and  $\lambda_i(\overline{\mathbf{A}}(\mathbf{X})) \rightarrow \lambda_i(\overline{\mathbf{A}}(\mathbf{X} + \Delta\mathbf{X}))$ .
- Closed-loop stability measure (Li 1998)

$$f(\mathbf{X}) \triangleq \min_{i \in \{1, \dots, m+n\}} \frac{1 - |\lambda_i(\overline{\mathbf{A}}(\mathbf{X}))|}{\sqrt{N} \left\| \frac{\partial \lambda_i(\overline{\mathbf{A}}(\mathbf{X}))}{\partial \mathbf{X}} \right\|_F}$$

where  $N = (l+n)(q+n)$  and  $\|\cdot\|_F$  the Frobenius norm.

- $f(\mathbf{X})$  quantifies the “robustness” of closed-loop stability for the realization  $\mathbf{X}$  to FWL controller perturbations:
- Under some mild conditions, the larger  $f(\mathbf{X})$ , the larger the FWL error  $\Delta\mathbf{X}$  that controller  $\mathbf{X}$  can tolerate without causing closed-loop instability.

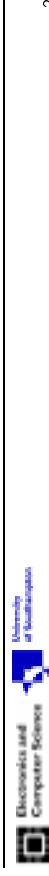


## Motivations and Background

Finite precision controller implementation can seriously influence closed-loop performance.

- Two types of finite word length errors: roundoff errors in arithmetic operations – controller signal errors, and controller coefficient representation errors – controller parameter errors.
- This work is concerned with FWL controller parameter errors, which have critical influence on closed-loop stability.

- Two strategies: direct and indirect.
- This work adopts an indirect approach.
- We present a novel search algorithm for global solutions to an existing optimal finite precision controller realization problem.



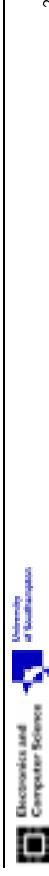
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## Optimal Realization Problem

- The optimal FWL controller realization problem (Li 1998)

$$v \stackrel{\Delta}{=} \max_{\mathbf{X} \in \mathcal{S}_C} f(\mathbf{X})$$

- Closed-form solutions was attempted in (Li 1998), but ended with suboptimal solutions.

- Direct numerical optimization has to apply: computationally costly and no way to know if a solution obtained is a global optimal solution.

- Our two-stage approach:

- Construct a closed-form realization set that contains global optimal solutions under a mild condition;
- Search in this set for a global solution with a numerical algorithm that are much more efficient than usual numerical optimization.

## Remarks

- The assumption that *Condition #* can be met is a reasonable one.

- It turns out that the set

$$\mathcal{X} \stackrel{\Delta}{=} \{\mathbf{X} : g(\mathbf{X}, k_1) = \rho_{k_1}, \mathbf{X} \in \mathcal{S}_C\}$$

can be constructed in a closed-form. Since  $\mathbf{X} = \mathbf{X}(\mathbf{T})$ ,  $\mathcal{X}$  is defined on the transformation set

$$\mathcal{T} \stackrel{\Delta}{=} \{\mathbf{T} : g(\mathbf{X}(\mathbf{T}), k_1) = \rho_{k_1}, \mathbf{T} \in \mathcal{R}^{n \times n}, \det \mathbf{T} \neq 0\}$$

- $\mathcal{T}$  can be searched for an  $\mathbf{T}_{\text{opt}}$  meeting

$$g(\mathbf{X}(\mathbf{T}_{\text{opt}}), k) \geq \rho_{k_1}, \quad \forall k \in \{1, \dots, m+n\} \setminus \{k_1\}$$

- This is much more efficient than minimizing  $f(\mathbf{X})$  over  $\mathcal{S}_C$ . Moreover, if a solution can be found in this way it is a global optimum for sure.

## Single-Pole Stability Functions and Their Peaks

- $\forall k \in \{1, \dots, m+n\}$ , define the single-pole FWL stability function of  $\mathbf{X}$

$$g(\mathbf{X}, k) \stackrel{\Delta}{=} \frac{1 - |\lambda_k(\overline{\mathbf{A}}(\mathbf{X}))|}{\sqrt{N} \left\| \frac{\partial \lambda_k(\overline{\mathbf{A}}(\mathbf{X}))}{\partial \mathbf{X}} \right\|_F}$$

and further define the single-pole peak of FWL stability as

$$\rho_k \stackrel{\Delta}{=} \max_{\mathbf{X} \in \mathcal{S}_C} g(\mathbf{X}, k)$$

- Theorem 1**  $v = \rho_{k_1} \stackrel{\Delta}{=} \min_{k \in \{1, \dots, m+n\}} \rho_k$  if and only if there exists  $\mathbf{X}_{\text{opt}} \in \mathcal{S}_C$  and  $k_1 \in \{1, \dots, m+n\}$  such that  $g(\mathbf{X}_{\text{opt}}, k_1) = \rho_{k_1}$  and

*Condition #*:  $g(\mathbf{X}_{\text{opt}}, k) \geq \rho_{k_1}, \forall k \in \{1, \dots, m+n\} \setminus \{k_1\}$

Obviously, such an  $\mathbf{X}_{\text{opt}}$  is a global optimal solution.

## Closed-Form Transformation Set

- Let  $\mathbf{p}_k$  and  $\mathbf{y}_k$  be the right and reciprocal left eigenvectors related to the closed-loop eigenvalue  $\lambda_k$ , respectively.

- Theorem 2** The value of  $\rho_k$  is easily determined, and  $g(\mathbf{X}(\mathbf{T}), k)$  achieves the maximum  $\rho_k$  if and only if

$$\mathbf{T} = \mathbf{Q} \begin{bmatrix} \mathbf{H}^{1/2} & \mathbf{0} \\ \mathbf{F}(\mathbf{H}^{1/2})^{-T} & \mathbf{\Omega} \end{bmatrix} \mathbf{V}$$

where  $\mathbf{V} \in \mathcal{R}^{n \times n}$  is an arbitrary orthogonal matrix, the orthogonal matrix  $\mathbf{Q}$  is known, and

- complex  $\mathbf{p}_k$  and  $\mathbf{y}_k$ :  $\mathbf{\Omega} \in \mathcal{R}^{(n-2) \times (n-2)}$  is an arbitrary nonsingular matrix, the  $2 \times 2$  matrix  $\mathbf{H}$  and the  $(n-2) \times 2$  matrix  $\mathbf{F}$  are known;
- real  $\mathbf{p}_k$  and  $\mathbf{y}_k$ :  $\mathbf{\Omega} \in \mathcal{R}^{(n-1) \times (n-1)}$  is an arbitrary nonsingular matrix, the scalar  $\mathbf{H}^{1/2} = \sqrt{h}$  and the  $(n-1) \times 1$  vector  $\mathbf{F}$  are known.

## Search Algorithm

- According to **Theorem 2** and setting  $\mathbf{V} = \mathbf{I}$ ,  $\mathcal{T}$  is given in the form:

$$\mathcal{T} = \left\{ \mathbf{T}(\Omega) : \mathbf{T}(\Omega) = \mathbf{Q} \begin{bmatrix} \mathbf{H}^{1/2} & \mathbf{0} \\ \mathbf{F}(\mathbf{H}^{1/2})^{-T} & \Omega \end{bmatrix} \right\}$$

- The objective: search for a nonsingular  $\Omega_{\text{opt}} \in \mathcal{R}^{(n-2) \times (n-2)}$  such that

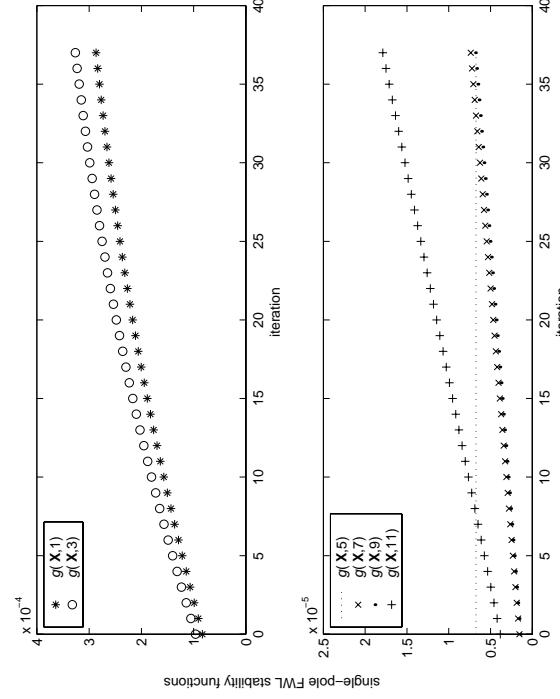
$$g(\mathbf{X}(\mathbf{T}(\Omega_{\text{opt}})), k) \geq \rho_{k_1}, \quad \forall k \in \{1, \dots, m+n\} \setminus \{k_1\}$$

(If  $\lambda_{k_1}$  is real-valued,  $\Omega \in \mathcal{R}^{(n-1) \times (n-1)}$ .)

- Notice that  $g(\mathbf{X}(\mathbf{T}(\Omega)), k)$  is differentiable with respect to  $\Omega$ .

- With the derivative, we know how to change  $\Omega$  so that  $g(\mathbf{X}, k)$  increase for those  $g(\mathbf{X}, k) < \rho_{k_1}$ , and  $g(\mathbf{X}, k)$  do not decrease for those  $g(\mathbf{X}, k) \geq \rho_{k_1}$   
 $\Rightarrow \Omega$  is updated iteratively until all the  $g(\mathbf{X}, k) \geq \rho_{k_1}$ .

## Iterative Search



## Design Example

- Li (1998):  $m = 5, n = 6, l = q = 1$ ; a closed-loop system of order 11.

- The controller transfer function  $C(z)$  has been given with the companion canonical form of  $C(z)$  as the initial realization.

- The closed-loop system: five pairs conjugate complex-valued eigenvalues  $\lambda_{1,2}, \lambda_{3,4}, \lambda_{5,6}, \lambda_{7,8}$  and  $\lambda_{9,10}$ , and one real-valued eigenvalue  $\lambda_{11}$ .

- The single-pole peaks of FWL stability are

$$\begin{aligned} \rho_{1,2} &= 2.5072e - 3, & \rho_{3,4} &= 2.1295e - 3, & \rho_{5,6} &= 6.7344e - 6, \\ \rho_{7,8} &= 2.8586e - 3, & \rho_{9,10} &= 3.0832e - 3, & \rho_{11} &= 4.3181e - 3. \end{aligned}$$

- The minimum value of all the  $\rho_k$ 's is  $\rho_5$  (or  $\rho_6$ )  $\Rightarrow k_1 = 5$  and the matrices  $\mathbf{Q}, \mathbf{H}$  and  $\mathbf{F}$  are determined  $\Rightarrow \mathcal{T} = \{\mathbf{T}(\Omega)\}$ .

## Global Optimal Solution

- During each iteration,  $\mathbf{X}(\mathbf{T}(\Omega))$  meets: 1)  $g(\mathbf{X}, k)$  increase for those  $g(\mathbf{X}, k) < \rho_{k_1}$ ; 2)  $g(\mathbf{X}, k)$  do not decrease for those  $g(\mathbf{X}, k) \geq \rho_{k_1}$ .

- At the 37th iteration, a global optimal realization  $\mathbf{X}(\mathbf{T}(\Omega_{\text{opt}}))$  is found, since at this stage Condition # is met:

$$g(\mathbf{X}(\mathbf{T}(\Omega_{\text{opt}})), k) \geq \rho_{k_1}, \quad k \in \{1, 2, \dots, 11\} \setminus \{5, 6\}$$

and the search algorithm terminated.

- Values of the closed-loop stability measure for the initial and global optimal realizations  $\mathbf{X}_0$  and  $\mathbf{X}(\mathbf{T}_{\text{opt}})$  are:

$$f(\mathbf{X}_0) = 3.1797 \times 10^{-11} \quad f(\mathbf{X}(\mathbf{T}_{\text{opt}})) = 6.7344 \times 10^{-6}$$

a factor of  $2 \times 10^5$  improvement in the closed-loop FWL stability measure.

## Conclusions and Future Works

- We have developed an efficient method to solve the optimal controller realization problem based on maximizing a closed-loop FWL stability measure.
  1. This method does not suffer from the drawbacks associated with using direct numerical optimization methods to tackle the problem.
  2. under a reasonable and mild condition, our method can find global optimal controller realizations for most practical systems.
- The arbitrary orthogonal matrix  $\mathbf{V} \in \mathcal{R}^{n \times n}$  in the closed-form transformation set  $\mathcal{T}$  can be explored to design:
  1. global optimal (in closed-loop stability sense) realizations of fixed-point controller with the smallest dynamic range.
  2. sparse global optimal realizations of fixed-point controller.