Hollow-core NANF for High-speed Short-reach Transmission in the S+C+L-bands

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Abstract-Hollow-core fibers offer a range of beneficial properties for high-speed optical communication applications, including low latency, low chromatic dispersion and nonlinearity. Based on a system employing a ~5-km span of hollow-core nested antiresonant nodeless fiber (NANF), we demonstrate in this paper, penalty-free Nyquist 4-ary pulse amplitude modulation transmission relative to the back-to-back at 80, 100 and 112 Gb/s over the S+C+L-bands. Furthermore, using multi-carrier directdetection optical orthogonal frequency division multiplexing, we show that compared to a standard single-mode fiber of a similar length, the NANF exhibits a significantly better bit error rate (BER) performance at 1550 nm and offers more than 20% capacity enhancement. Consequently, up to 138.09-Gb/s capacity has been achieved at a BER threshold of 3.8×10⁻³ after ~5-km transmission. We conclude that the use of NANF offers a promising solution for achieving simultaneously high capacity and low latency in future short-reach optical interconnects.

Index Terms—Hollow-core Fibers; Ultra-wideband Transmission; Intensity-modulation and Direct-detection; Short-reach Optical Interconnects.

I. INTRODUCTION

TOLLOW-CORE optical fibers based on nested antiresonant nodeless fiber (NANF) technology have emerged recently as a promising evolution from conventional standard single mode fiber (SSMF) for optical data transmission [1-2]. NANFs can be designed to offer low-loss and effectively single-mode propagation over an extended optical bandwidth [3-4]. Since they rely on guidance in a hollow core, they benefit from propagation speeds that approach the speed of light in vacuum, thereby offering approximately 30% lower latency in transmission than SSMFs [2]. Furthermore, hollow-core guidance also results in much more linear optical fiber transmission, thereby avoiding the onset of deleterious nonlinear effects. These benefits offer the potential to revolutionize the design of optical networks once multikilometer long fiber cables exhibiting similar/lower losses than SSMF can be robustly manufactured and deployed. Moreover,

an additional benefit of hollow-core guidance in NANFs is that the fiber chromatic dispersion (CD) takes extremely low values over an extended wavelength region away from the edges of its passband [5]. While the loss of NANFs has been reduced rapidly over the last few years [6-8], the low CD of NANF promises to have more immediate impact in shorter-reach optical systems, which are not loss-limited. For example, the low CD may alleviate the power fading issue in intensity-modulation/direct-detection (IM/DD) systems and also allow for the use of less complex equalization for CD compensation/mitigation in coherent transmission.

In this context, this paper considers short-reach (<10 km long) optical links, which play a vital role in today's optical networks, since they are the dominant contributors to data traffic [9-11]. Of increasing importance are connections between edge data centers, edge nodes and producers/consumers [12] which support rapidly expanding traffic demand and enable low-latency applications in today's networks. Driven by this, the realization of high-speed shortreach optical interconnects has drawn significant research interest from both academia and industry in recent years [13-14]. To date, intensity-modulation/direct-detection (IM/DD) is the dominant solution for short-reach interconnects with link lengths ranging from a few meters to a few kilometers. This is primarily because of the inherent advantages of IM/DD in terms of simplicity, cost, power consumption and footprint over coherent approaches [14-15]. Moving towards standardization of 400G systems, IM/DD is still prioritized for such short-reach applications [16-17], and the IEEE has established the P802.3bm Task Force to address this demand [18]. Meanwhile, it is worth noting that the use of coherent detection in short-reach access is also currently being considered, and is included in the 400ZR implementation agreement by the Optical Internetworking Forum [19]. While there have already been some demonstrations that highlight the NANF's properties which would be greatly beneficial to highspeed coherent transmission systems [20-21], in this work, our focus lies in the use of NANF for IM/DD applications.

In the literature, research and development efforts have

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tackled different aspects of high-speed IM/DD transmission, including transceivers, advanced digital signal processing (DSP) and modulation formats [22-28]. With respect to the latter, 4-ary pulse amplitude modulation (PAM4), carrierless amplitude and phase (CAP) modulation, and DD optical orthogonal frequency division multiplexing (DDO-OFDM) with adaptive bit-and-power loading (also known as discrete multitone (DMT) modulation), have been extensively investigated [27-28]. An experimental comparison amongst different modulation formats in 100-Gb/s/λ short-reach interconnects was reported in [28], showing their respective advantages in terms of receiver sensitivity, spectral efficiency, and low computational complexity. However, it is challenging to realize high-speed C-band transmission over SSMF with IM/DD technology, an important performance-limiting factor being CD-induced power fading [29]. For example, it is straightforward to show that the spectrum of broadband signals will experience a first spectral null at ~25 GHz after transmission over 5-km of SSMF at 1550 nm. Generally, CD effects in short-reach SSMF links are mitigated either by using spectrally-efficient modulation formats in C-band transmission systems [28-29] or by adopting the O-band for data transmission. However, the O-band is highly susceptible to nonlinearities of SSMFs even for short-reach links [30] due to the low CD in this spectral region, which restricts the number of high-speed wavelength-division multiplexed (WDM) channels. In contrast, hollow-core NANFs are effectively nonlinearity-free [21] and exhibit ultra-low CD over a broad bandwidth, thus offering the potential to achieve extremely high capacities using wideband densely-spaced WDM transmission.

In this work, we study IM/DD systems that use either Nyquist PAM4 or adaptively-loaded DDO-OFDM, and are constructed based on a ~5-km NANF span. We demonstrate that relative to the back-to-back (B2B), NANF allows for penalty-free Nyquist PAM4 transmission over the S+C+L-bands at speeds of 80, 100

and 112 Gb/s after transmission in \sim 5 km of this fiber. This performance significantly outperforms that of a similar length of SSMF. Using the adaptively-loaded DDO-OFDM, we further show that >20% capacity improvement over a SSMF of a similar length can be achieved in NANF, and up to 138.09-Gb/s IM/DD transmission can be realized at a bit error rate (BER) threshold of 3.8×10^{-3} .

The rest of this paper is organized as follows. Section II presents a numerical comparison between SSMF and NANF. Section III shows the detailed experimental setup used in our transmission experiments. The results of Nyquist PAM4 transmission are given in Section IV whilst the results of DDO-OFDM are shown in Section V. Finally, Section VI concludes this work. This paper expands on the results that were originally presented in [5].

II. NUMERICAL SIMULATIONS

Fig. 1 shows the results of a numerical comparison between SSMF (ITU G.652 [31]) and a NANF (whose design has been detailed in [32]) of 5-km length. For reference, we also show the results obtained for a non-zero dispersion-shifted fiber (NZ-DSF, ITU G.655 [33]), as an additional low-CD solid-core fiber alternative for the C-band. We first compared the simulated CD of the fibers over the S+C+L-bands. The SSMF and NZ-DSF calculations assumed the mean values of the CD range, as specified in the ITU standards for the wavelengths of interest [31, 33]. It is seen in Fig. 1(a) that the NANF has a much lower CD as well as a much lower CD slope than SSMF across the 1.4 to 1.625-µm wavelength range. It is worth noting that the CD is directly responsible for power fading in IM/DD, which imposes a limit on the usable bandwidth at a given wavelength and transmission distance. An example of this is shown in Fig. 1(b), which compares the frequency response of the three fiber types at a wavelength of 1550 nm and over a distance of 5 km. The CD values of the SSMF, NZ-DSF and NANF at this wavelength

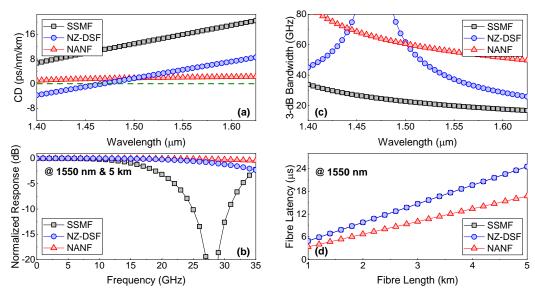


Fig. 1. Numerical comparisons amongst the SSMF, NZ-DSF and NANF at a length of 5 km: (a) CD, (b) 3-dB bandwidth, (c) profiles of CD-induced power fading after 5-km 1550-nm transmission and (d) fibre latency at 1550 nm. Note that the results of SSMF and NZ-DSF overlapped with each other in Fig. 1(d).

have been assumed to be 15.95, 4.50 and 2.04 ps/nm/km, respectively. The 3-dB bandwidth of the three fiber types is compared in Fig. 1(c) across the 1.4 to 1.625-µm wavelength range. It is observed that the NANF exhibits more than twice the bandwidth across this entire wavelength range compared to the SSMF. It also offers a wider bandwidth than the NZ-DSF, apart from within a ~60-nm range centered around 1460 nm, where the CD of the NZ-DSF approaches zero. We note that although the bandwidth of the NZ-DSF is broader than that of the NANF in this ~60 nm window, the available bandwidth in the NANF (>60 GHz) is already well above the bandwidth requirements of 400/800G standards [16-18]. Furthermore, this graph does not account for the effects of fiber nonlinearities, which are more pronounced in the low-CD region of the NZ-DSF (i.e. where its bandwidth is the broadest), and which in turn will restrict the fiber's capability for wavelength-division multiplexing transmission [34]. In contrast, the NANF is effectively free of nonlinearities. We also note that discussion of fiber loss was excluded from the numerical simulations when evaluating the CD effect in the three types of fibers. The impact of fiber loss on transmission, which is also an important parameter in this work, will be experimentally evaluated in Section IV. We further investigated the latency performance of the different fibers, using (effective) refractive indices of 1.4682 (Corning SMF28e), 1.4693 (Corning LEAF), and 1.003 [2] for the SSMF, NZ-DSF and NANF, respectively. The results are shown in Fig. 1(d). It is seen that even at a link length of just 5 km, the NANF offers a ~8-us latency reduction compared to the two solid-core fibers.

III. EXPERIMENTAL SETUP

A. System configuration

To evaluate the transmission performance in SSMF and NANF, we implemented an IM/DD transmission system, as shown in Fig. 2 (note that the case of NZ-DSF was not considered, since a suitable length of NZ-DSF was not available while performing these experiments). At the transmitter, a tunable laser capable of generating an optical carrier over the range from 1460 nm to 1640 nm was connected to a Mach-Zehnder modulator (MZM), which was driven by an electrically amplified data signal generated by an arbitrary waveform generator (AWG) followed by an electrical amplifier (EA). The output of the MZM was directly launched into the transmission link (either NANF or SSMF) without optical amplification. In addition, we have also considered the B2B case, as indicated in Fig. 2. In the NANF case, two structurally matched NANFs were directly spliced together to produce a spool of ~4.76 km length. We note that similar to the splicing of SSMFs, the NANFs were arc-fusion spliced after having been structurally aligned with each other, and the splice loss was around 0.1 dB. The average core size and membrane thickness of the first ~3.4km length of NANF (#1) were 36.4 µm and 0.55 µm, respectively, whilst the corresponding values of the second ~1.4-km length of NANF (#2) were 34.6 μ m and 0.50 μ m, respectively [32]. Scanning electron microscope (SEM) images of the two NANFs are shown in Fig. 2, showing symmetric fiber cross-sections. The two ends of the spliced NANF were connectorized through SSMF fiber pigtails, and these multiple splices contributed to around 2 dB loss. The total loss of the SSMF-connectorized NANF ranged from 6.45 to 8.33 dB over the S+C+L-bands, as will be detailed in Section IV below. In the SSMF case, the total fiber length was ~4.53 km, which comprised two spools of ~4-km and ~0.5 km lengths, respectively.

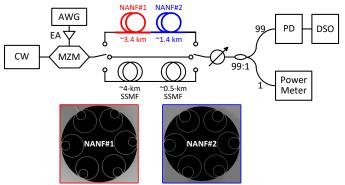


Fig. 2. Experimental setup of the IM/DD system. Insets: SEM images of the used NANFs.

At the receiver, an optical attenuator was used to vary the received optical power (ROP) at the photodetector (PD), which was monitored using a 99:1 optical coupler. The detected electrical signal at the PD was then recorded by an 80-GSa/s digital storage oscilloscope (DSO) for further offline DSP.

B. DSP blocks

Different DSP processes were required aforementioned two modulation formats in this work. For the generation of the single-carrier Nyquist PAM4 (Fig. 3(a)), the PAM4-mapped signal was first 70 times up-sampled and then Nyquist-filtered using a square-root raised cosine (SRRC) filter. The roll-off factor of the SRRC filter was 0.1. Then, the filtered signal was 40-/50-/56-times down-sampled and digitalto-analog converted in the AWG operating at 70 GSa/s, to 80-/100-/112-Gb/s Nyquist PAM4 respectively. At the receiver side, the synchronized data was resampled and match-filtered before being fed to a T/2-spaced (17,7)-tap decision feedback equalizer (DFE). The recursive least squares (RLS) algorithm was used in the DFE and its forgetting factor was 0.998. Finally, the equalized signal was PAM4 de-mapped and the corresponding BER was calculated via error counting.

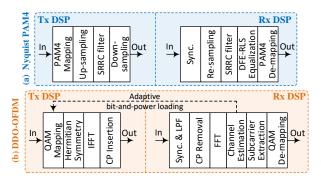


Fig. 3. Block diagrams of (a) the Nyquist PAM4 and (b) the DDO-OFDM.

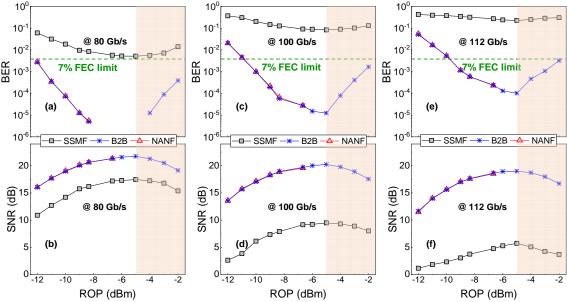


Fig. 4. Comparisons of BER and SNR performance versus ROP of 1550-nm Nyquist PAM4 transmission in SSMF, B2B and NANF links at (a)-(b): 80 Gb/s, (c)-(d): 100 Gb/s, and (e)-(f): 112 Gb/s. Shaded region: the limited power range of the transimpedance amplifier causes spurious results when ROP > -5 dBm.

For the DDO-OFDM transmitter (Fig. 3(b)), after quadrature amplitude modulation (QAM) mapping, Hermitian symmetry operation was implemented to ensure real-valued outputs after the 512-point inverse fast Fourier transform (IFFT). Before being sent to the 70-GSa/s AWG, a cyclic prefix with a length of 1/16 DDO-OFDM symbols was adopted. At the receiver, the recorded data was synchronized and low-pass filtered, followed by the CP removal. Then, simple one-tap equalization was performed, and only the data subcarriers were extracted for the subsequent QAM de-mapping. The corresponding BER was also obtained via error counting. For the adaptively-loaded DDO-OFDM, pilot transmission was conducted to estimate the channel state information, which was then utilized at the transmitter for adaptive bit-and-power loading [4, 35].

IV. SINGLE-CARRIER NYQUIST-PAM4 TRANSMISSION

In this section, we investigate the transmission performance of the Nyquist PAM4 signals in the B2B, SSMF and NANF links. To characterize the impact of CD on high-speed transmission, the data rates of 80, 100 and 112 Gb/s were considered in the experiments.

A. BER&SNR versus ROP at 1550 nm

We first studied the BER and signal-to-noise ratio (SNR) performance versus ROP at different speeds for the three links at the wavelength of 1550 nm. Fig. 4 shows the corresponding transmission results. It is seen that under all the investigated data rates (i.e., 80, 100 and 112 Gb/s), similar BERs and SNRs as for the B2B were achieved in the NANF, confirming that the fiber maintains the integrity of the signals over this distance. In both of these cases, the required ROPs to obtain a BER below 3.8×10^{-3} , i.e., the 7% forward error correction (FEC) limit, were around -12 dBm, -11 dBm, and -10 dBm, when transmitting at 80-Gb/s, 100-Gb/s and 112-Gb/s, respectively. In contrast, regardless of the ROP, the BER measured when transmitting in

the SSMF was always above the 7% FEC limit, even for the case of 80-Gb/s. Indeed, an 80-Gb/s PAM4 signal has a 3-dB bandwidth of ~20 GHz, and Fig. 1(b) suggests that it will be shaped substantially due to power fading in the dispersive fiber. This effect is also reflected in the SNR plots of Figs. 4(b), 4(d) and 4(f), which confirm that transmission at 80-Gb/s, 100-Gb/s and 112-Gb/s over the SSMF suffers an additional penalty of around 5 dB, 11 dB and 14 dB, respectively, relative to the NANF (and B2B).

It is noted that when increasing the ROP to values beyond -5 dBm, degraded BER and SNR performance was experienced in all cases, as shown in the shaded region in Fig. 4. This was because of the limited input power range of the transimpedance amplifier in the PD. Therefore, in the following experiments, to avoid the high-ROP-induced performance degradation, an upper ROP limit of -5 dBm was imposed on the PD.

B. BER&SNR performance over S+C+L-bands

We next tuned the laser across the S+C+L-bands to investigate the transmission performance in B2B/SSMF/NANF links. While doing this, the total loss of the ~5-km SSMF-connectorized NANF over the S+C+L-bands was experimentally characterized, and the corresponding results are shown in Fig. 5.

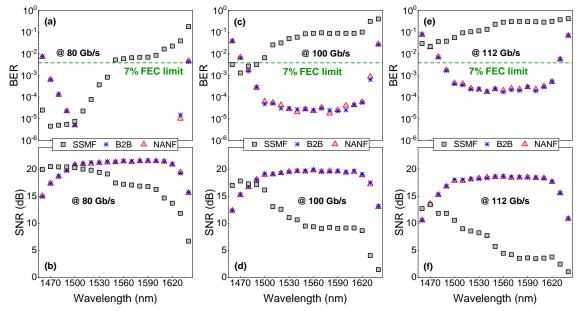


Fig. 6. Comparisons of BER and SNR performance across the S+C+L-bands of Nyquist PAM4 transmission in SSMF, B2B and NANF links at (a)-(b): 80 Gb/s, (c)-(d): 100 Gb/s, and (e)-(f): 112 Gb/s.

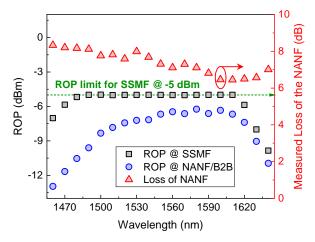


Fig. 5. ROP in different links and the measured loss of the SSMF-connectorized NANF over S+C+L-bands.

The figure shows that the insertion loss of the NANF link ranged from 6.45 to 8.33 dB in the spectral region of interest. Note that these values include the aforementioned ~2-dB splicing losses. A minimum loss of 6.45 dB was measured at 1610 nm.

The transmission performance at the three data rates was then measured across this entire wavelength range. This was evaluated at the maximum ROP that was available, and which is also plotted in Fig.5. Note that in order to ensure a meaningful comparison with the B2B case, the ROPs for the B2B link were maintained at the same values as in the NANF case. In the case of the SSMF, the ROP was capped at -5 dBm, to avoid nonlinear distortions at the PD, as mentioned above. The reduced ROPs at the two edges of the spectral region were due to the limited power provided by our tunable laser at these wavelengths.

The results of the 80-Gb/s, 100-Gb/s and 112-Gb/s Nyquist PAM4 transmission across the S+C+L-bands are shown in Fig. 6. It is evident that similar BER and SNR values can be

achieved over the investigated spectral region at all these data rates for the NANF as in the B2B, validating the capability of the NANF for penalty-free high-speed transmission over an ultra-wide bandwidth after around 5 km. Across all C-band wavelengths and most of the S- and L-band wavelengths, a BER lower than the 7% FEC limit can be achieved in the NANF even at the highest data rate of 112 Gb/s that we considered. In contrast, only a few wavelengths at the short-wavelength edge of the S-band can marginally achieve a BER below 3.8×10⁻³ in the SSMF at data rates above 100 Gb/s. At the lower data rate of 80 Gb/s, acceptable BER performance (<3.8×10⁻³) was achieved at wavelengths shorter than 1550 nm, where the impact of CD was less severe. Note that it is only at the shorter wavelengths of the S-band, where the higher ROP that was available in the SSMF transmission offered an edge in the performance relative to the NANF, especially for the 80-Gb/s case (Figs. 6(a) and 6(b)). This resulted both from the higher loss of the NANF, and the diminishing CD values in the SSMF at these wavelengths.

It is also worth noting that the signal performance was largely dependent on the wavelength while transmitting in SSMF (reflecting the substantial changes in the CD values across this wide spectral range), which is in contrast to the transmission in the NANF, which exhibits almost uniform and low CD values across this entire spectral range (see Fig. 1(a)).

V. MULTI-CARRIER DDO-OFDM TRANSMISSION

We next adopted multi-carrier DDO-OFDM as the modulation format, since its signal bandwidth can be flexibly adjusted by altering the number of data subcarriers, transmission capacity can also be maximized via adaptive bit-and-power loading. Since the previous Section ascertained that transmission in the NANF over the S+C+L-bands after ~5 km is penalty-free relative to the B2B link, we carried out measurements in this Section at one wavelength only, namely

1550 nm. We started by comparing the performance of uniformly-loaded DDO-OFDM in the NANF and the SSMF, and then studied the transmission of adaptively-loaded DDO-OFDM in the two fiber types.

A. Performance of the uniformly-loaded DDO-OFDM

Fig. 7 shows a comparison of the BER and average SNR performance of uniformly-loaded DDO-OFDM transmission in SSMF and NANF, for a varying number of subcarriers and a fixed ROP of -7 dBm. Here, the average SNR refers to the mean value of the SNRs of all data subcarriers. All the data subcarriers were loaded with 16QAM signals and no power loading was implemented. Therefore, when changing the number of data subcarriers from 100 to 240, the corresponding data rate varied from 54.6875 Gb/s to 131.25 Gb/s. Note that the maximum number of data subcarriers, namely 240, was limited by the cut-off frequency of the DSO. It is seen that the NANF offered a lower BER and a higher average SNR than the SSMF in all cases, yet the differences in the performance were less pronounced when the number of data subcarriers (hence, the signal bandwidth) was small. In accordance to the discussions in the preceding Sections, this is because the extent of the CD-induced power fading depends on the signal bandwidth.

It is worth noting that with an increase in the number of data subcarriers, i.e. the signal bandwidth, the BER and SNR performance of the NANF transmission also degraded, as shown in Fig. 7. This is attributed to the roll-off effect of the bandwidth-limited transceiver, rather than the NANF itself.

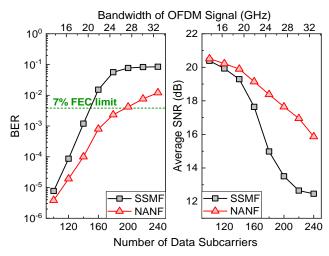


Fig. 7. BER and average SNR versus number of data subcarriers for uniformly-loaded DDO-OFDM transmission in ~5-km SSMF and NANF. The ROPs in all cases were fixed to -7 dBm.

The SNR profiles of the uniformly-loaded 16QAM-DDO-OFDM with 240 data subcarriers after transmission in the NANF and the SSMF are shown in Fig. 8. It is clear that the transmission in the NANF did not suffer from power fading, whereas in the case of the SSMF, a severe spectral null could be observed around 24 GHz. Note that the two low-SNR spikes shown in both Figs. 8(a) and 8(b) were caused by clock leakage of the high-speed AWG rather than any of the fibers.

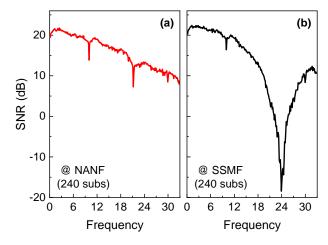


Fig. 8. SNR profiles of uniformly-loaded DDO-OFDM after ~5-km transmission in (a) NANF and (b) SSMF. The ROP in both cases was -7 dBm.

Using 240 data subcarriers in the case of uniformly-loaded DDO-OFDM, i.e. a signal bandwidth of 32.8125 GHz, we further varied the ROP at the PD and compared the corresponding performance in the NANF and the SSMF. To obtain acceptable BER performance, all data subcarriers were loaded with 4QAM signals, hence the data rate was reduced relative to the experiments of Fig.7 to 65.625 Gb/s. The corresponding transmission results are shown in Fig. 9. Due to the severe spectral fading within the signal bandwidth, the BERs of the SSMF transmission were always above the 7% FEC limit, regardless of the ROP. In comparison, significant BER performance improvement can be enabled when transmitting the same signal in the NANF. A BER lower than the 7% FEC limit was achieved for a ROP greater than -10 dBm. Accordingly, the NANF exhibited around 4.5-dB improvement over the SSMF in terms of the average SNR performance.

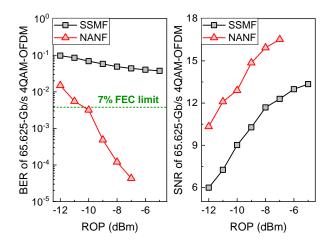


Fig. 9. BER and average SNR versus ROP for 65.625-Gb/s uniformly-loaded 4QAM-DDO-OFDM transmission in ~5-km SSMF and NANF.

B. Performance of the adaptively-loaded DDO-OFDM

The final set of transmission tests used adaptively-loaded DDO-OFDM, validating that the NANF can offer significant performance improvement even when the adaptive bit-and-power loading is employed.

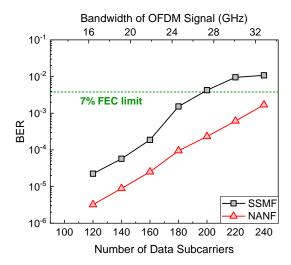


Fig. 10. BER versus number of data subcarriers for adaptively-loaded DDO-OFDM transmission in \sim 5-km SSMF and NANF. The ROPs in all cases were fixed to -7 dBm.

Fig. 10 shows the BER performance of the adaptively-loaded DDO-OFDM transmission as a function of the number of data subcarriers, for a fixed ROP of -7 dBm. To show a fair comparison with the results shown in Fig. 7, the target of the adaptive bit-and-power loading applied here was to optimize the BER performance whilst maintaining the same data rates as those in Fig. 7, i.e. keeping an identical spectral efficiency of 4 bit/symbol/Hz. Furthermore, since some subcarriers may not be loaded with data, the comparison of average SNR performance is omitted. As anticipated, the figure shows that the BER performance of both SSMF and NANF transmission improved through the use of adaptive bit-and-power loading. As a result, more than 240 subcarriers, i.e. >131.25 Gb/s total capacity, could be supported in the NANF while maintaining a BER below the 7% FEC limit. The maximum capacity of the SSMF transmission system was ~109.375 Gb/s (~200 subcarriers) under the same BER threshold.

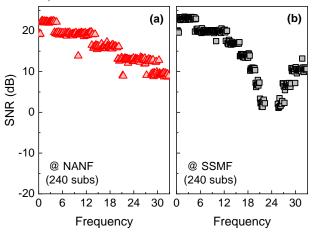


Fig. 11. SNR profiles of adaptively-loaded DDO-OFDM after ~5-km transmission in (a) NANF and (b) SSMF. The ROP in both cases was -7 dBm.

The corresponding SNR profiles of the 240-subcarrier adaptively-loaded DDO-OFDM transmission in the NANF and the SSMF are shown in Fig. 11. Clearly, the profiles generally follow the patterns shown in Fig. 8. The difference is that the subcarriers loaded with the same order of QAM formats shared

comparable SNRs and those under extremely low SNR conditions were not data-loaded. This is the case for the subcarriers falling around the spectral null region of the SSMF. In comparison, the SNRs of the subcarriers in the NANF case exhibited a staircase-like profile, since the transmission did not suffer from severe power fading thanks to the ultra-low CD of the fiber.

Finally, we investigated the maximum achievable capacity of the adaptively-loaded DDO-OFDM transmission in the two fiber types under different ROPs, where the objective of the adaptive bit-and-power loading was to maximize the transmission capacity while satisfying the 7% FEC limit. A signal bandwidth of 32.8125 GHz, i.e., 240 data subcarriers, was adopted, and the corresponding results are shown in Fig. 12. At all ROPs, around 20% to 30% capacity improvement could be achieved by using the NANF in the transmission relative to the SSMF. After ~5-km transmission, at a ROP of 7 dBm, the SSMF enabled a ~112.52-Gb/s capacity. In comparison, under the same conditions, the NANF could support up to ~138.09 Gb/s.

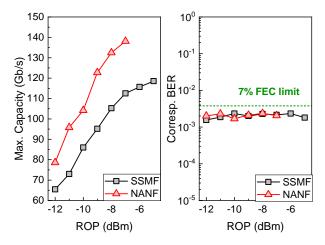


Fig. 12. Comparison of the maximum achievable capacity versus ROP in SSMF and NANF, and the corresponding BER.

VI. CONCLUSION

In this paper, we have presented a comprehensive experimental study of IM/DD transmission performance in a 5km length of NANF over the S+C+L-bands. We performed comparative transmission tests over a similar length of SSMF and confirmed that the NANF allows for penalty-free Nyquist PAM4 transmission at speeds of 80, 100 and 112 Gb/s across this entire wavelength range over a length of ~5 km. This was contrasted to the performance in the SSMF, which was significantly dependent on the signal wavelength and could not support signals with a higher data rate than 80Gb/s for a large proportion of these wavelengths over this distance. We anticipate that the performance gap between the NANF and SSMF in terms of the achievable capacity will further increase with an increase in transmission reach, thanks to the low CD of the NANF. We also adopted the DDO-OFDM format for further performance comparison between the NANF and the SSMF at 1550 nm. We showed that >20% capacity enhancement could be achieved in the NANF relative to the

SSMF at this distance, while operating below the 7% FEC limit. Consequently, a capacity up to 138.09-Gb/s was achieved in the NANF at a ROP of -7 dBm for the adaptively-loaded DDO-OFDM transmission. Our results indicate that even though at their current stage of development, NANFs exhibit a higher propagation loss than SSMF, they already offer significantly better performance in terms of both achievable capacity and latency (~30% reduction) in short-reach IM-DD transmission systems, highlighting their potential for capacity-hungry and latency-sensitive applications like intra-data center links. We believe that the results presented in this work demonstrate a viable route towards fulfilling the increasing capacity and latency requirements of future optical transmission and mobile networks through the adoption of this novel fiber technology.

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