

Realisation of broadband two-dimensional nonreciprocal acoustics using an active acoustic metasurface

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1 Nonreciprocal acoustic devices have been shown to be able to control incident waves
2 propagating in one direction, whilst allowing incident waves propagating in the oppo-
3 site direction to be transmitted without modification. Nonreciprocal sound transmis-
4 sion has typically been achieved by introducing nonlinearities or directional biasing
5 through fluid motion or spatiotemporal modulation of resonant cavities. However, the
6 spatial arrangement of these approaches creates preferential characteristics in one di-
7 rection, such that the direction of the nonreciprocal behaviour is fixed and, thus, they
8 are not straightforwardly reconfigurable. To address this issue, it has previously been
9 shown that feedforward wave-based active controllers can be used to drive a single
10 subwavelength active unit cell to achieve broadband nonreciprocal sound transmission
11 or absorption in a one-dimensional linear acoustic system. Extending this concept,
12 this paper investigates how the feedforward wave-based active controller can be used
13 to drive an array of subwavelength active unit cells forming a metasurface to achieve
14 broadband nonreciprocal sound absorption over a two-dimensional plane. Through
15 both simulation and experimental studies, this paper shows that active wave-based
16 absorption control systems can achieve broadband nonreciprocal sound absorption
17 when the incident waves are generated by both normally and obliquely-positioned
18 primary sources.

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19 **I. INTRODUCTION**

20 Reciprocity is an acoustic property that describes the symmetry in sound transmission be-
21 tween two points. For example, the acoustic response between an acoustic source and sensor
22 is equal to the response when the acoustic source and sensor locations are swapped. Al-
23 though a variety of acoustic applications have exploited reciprocity to simplify measurement
24 processes^{1,2}, it is undesirable in certain applications. For example, nonreciprocal acoustic
25 control could be exploited variously to improve privacy, to acoustically cloak objects or
26 to enable improved sensing or manipulation of a sound field. This has led to significant
27 interest in the development of nonreciprocal acoustic devices that achieve one-way sound
28 transmission. Previously proposed nonreciprocal acoustic devices have generally broken the
29 symmetry in transmission by introducing nonlinearities³⁻⁷, fluid motion⁸ or spatiotemporal
30 modulation of resonant cavities⁹⁻¹¹. These various approaches have different limitations:
31 nonlinear nonreciprocal acoustic devices typically require high input power and are often
32 bulky; additional unwanted noise is introduced when fluid motion is introduced; resonant
33 cavities can only achieve nonreciprocal behaviour over a narrow bandwidth and finally, all of
34 these nonreciprocal devices are not straightforwardly tuneable to reverse the direction of the
35 nonreciprocal behaviour. The majority of these limitations have been addressed through the
36 development of non-local active metamaterials that contain non-located sensor and actua-
37 tor pairs^{12,13}. However, the spatial arrangement of the sensor and actuator pairs still creates
38 preferential characteristics in one direction and, thus, the direction of the nonreciprocal be-
39 haviour is still fixed in this case. More recently, the issue of tuneability has been addressed

40 through the development of feedforward wave-based active control systems that drive one
41 subwavelength active unit cell to minimise the transmitted and reflected wave components
42 individually to achieve broadband nonreciprocal sound transmission or absorption in a one-
43 dimensional linear acoustic system¹⁶. This wave-based active controller achieves broadband
44 nonreciprocal control by taking advantage of the causality of feedforward control. By set-
45 ting the positive propagating incident wave as the reference signal, this controller minimises
46 the corresponding wave components with respect to the positive propagating incident wave,
47 whilst the negative propagating incident wave propagates unimpeded since the reference
48 signal to the controller in this case is near-zero and the wave-based active controller only
49 controls the wave components that are associated with the reference signal. The advantage
50 of a wave-based active control system is that it is fully tuneable, such that the direction
51 of the nonreciprocal behaviour can be easily reversed by changing the reference and error
52 signals in the feedforward controller.

53 Since the study of nonreciprocal acoustic devices is a fairly new topic, there has been
54 limited investigation into how nonreciprocal behaviour can be achieved in two or three-
55 dimensional spaces^{17,18}. Building on the wave-based active control concept, this paper there-
56 fore presents an investigation into how the feedforward wave-based active controller proposed
57 in¹⁶ can be extended to form a metasurface, consisting of an array of control sources and
58 arrays of pressure sensors, which is used to minimise the transmitted and reflected wave com-
59 ponents to achieve broadband nonreciprocal sound absorption in a two-dimensional plane
60 within a three-dimensional space. This extension over the work presented in¹⁶ not only
61 includes extension to a multi-input, multi-output metasurface, but also includes the intro-

duction of a more complex wave-separation technique that is able to handle primary wave fields with spherical wave fronts and non-normal angles of incidence. Therefore, this paper presents two main research contributions: firstly, it is demonstrated how the wave-based active controller can be extended to drive an active metasurface to achieve nonreciprocal sound absorption across a two-dimensional plane and secondly, it is shown that the wave separation method can handle spherical wave fronts and obliquely-positioned primary sources. The performance of the proposed wave-based actively controlled metasurface has initially been investigated via simulations using an analytical free field model with monopole acoustic sources. The proposed methodology is then investigated using the responses measured for a practical system constructed in the anechoic chamber at the Institute of Sound and Vibration Research (ISVR).

II. SYSTEM DESCRIPTION

The physical arrangement of the proposed wave-based nonreciprocal actively controlled metasurface is shown in Figure 1. The system consists of a dual-layer array of monopole control sources, which are indicated by the crosses in Figure 1, and dual-layer arrays of pressure sensors positioned either side of the control source array. The dual layer array of sources is required to be able control the reflected and transmitted wave components; whilst the dual layer arrays of sensors are required to be able to sense the incident, reflected and transmitted wave components variously. The incident acoustic fields propagating in the positive and negative directions are generated by primary sources located either side of the control source and sensor arrays, as also shown in Figure 1. The acoustic field can be

83 described in terms of wave components, which are indicated by the coefficients A to D in
 84 Figure 1. It is worth noting that the coefficients A to D represent different wave components
 85 depending on the direction of the incident sound field. When the incident sound field is
 86 generated by a positive primary source, A and C are the positive propagating incident and
 87 transmitted waves respectively and B and D are the negative propagating upstream and
 88 downstream reflected wave components respectively. Conversely, when the incident sound
 89 field is produced by a negative primary source, D and B are the negative propagating
 90 incident and transmitted waves respectively and C and A are the positive propagating
 91 upstream and downstream reflected wave components respectively. The subscripts $+$ and $-$
 92 are used throughout this paper to distinguish whether a wave component is generated by
 93 a positive or negative primary source. Each pressure sensor in Figure 1 is denoted by the
 94 following notation: the first subscript indicates the sensor plane within which the sensor is
 95 located and the second subscript indicates which of the L -th pressure sensors within that
 96 sensor plane is being referred to. L is also the number of error signals, which will be described
 97 in Section IV.

98 The geometric parameters that define the system shown in Figure 1 are presented in Table
 99 I. The frequency range of interest for this investigation has been defined between 150 Hz
 100 and 400 Hz. The lower limit is based on the low frequency performance of the loudspeakers
 101 used to practically realise the control sources in the experimental implementation and the
 102 upper frequency limit has been chosen to avoid spatial aliasing by ensuring that the spacing
 103 between control sources and sensors in the y -direction is less than half the shortest acoustic
 104 wavelength. The frequency range could be extended by using loudspeakers with an improved

105 low frequency performance and reducing the spacing in the y -direction between adjacent
 106 sources and pressure sensors. However, these requirements are somewhat contradictory,
 107 since improving the low frequency performance of loudspeakers typically requires their size
 108 to increase which in turn will limit how close to each other the loudspeakers can be placed.

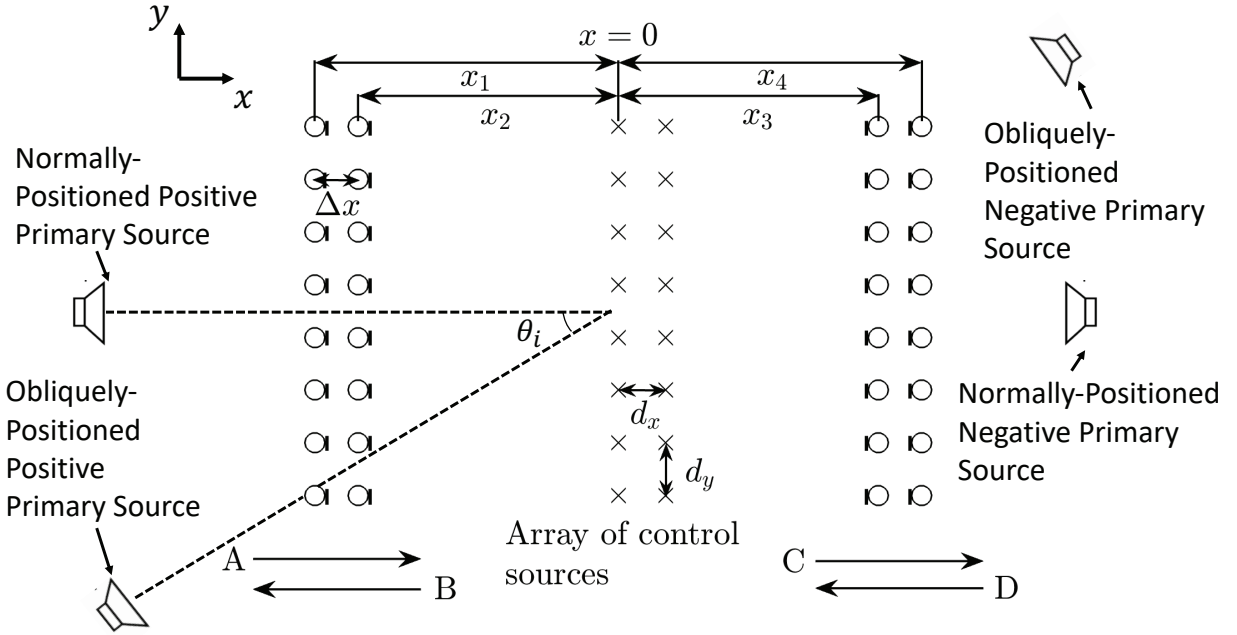


FIG. 1. The system geometry used to realise nonreciprocal sound absorption in a two-dimensional plane via the wave-based actively controlled metasurface. The black crosses indicate the locations of control sources and the circles denote the locations of the pressure sensors. The primary sources are denoted by loudspeaker diagrams, but their positions in the subsequent investigations are not shown to scale in this diagram.

109 It is important to note that we have assumed in this paper that we are considering control
 110 of the sound field in a two-dimensional plane within a three-dimensional space. This means
 111 that we are able to control the sound field using line arrays of sources and sensors, as shown

Variable	Value	Variable	Value
d_y	0.4 m	d_x	0.11 m
Δx	0.15 m	x_1	-0.725 m
x_2	-0.575 m	x_3	0.615 m
x_4	0.765 m		

TABLE I. The parameters used to define the system shown in Figure 1.

112 in Figure 1, rather than the grids of sources and sensors that would be required to control
 113 over three-dimensional space. The methodology described here could be extended to control
 114 over a full three-dimensional space, but in addition to grids of sources and sensors, would
 115 also require an extension of the wave separation method described in the next section to use
 116 a two-dimensional Spatial Fourier Transform. This is left for future work, mainly due to the
 117 practical challenge of realising such a large experimental implementation.

118 III. WAVE SEPARATION METHOD

119 The wave separation method described in this section focuses primarily on the case when
 120 the incident wave is generated by a positive primary source, as shown in Figure 1, however,
 121 the same methodology can also be used in the case of a negative primary source. As in¹⁶,
 122 the positive propagating incident, A_+ , transmitted, C_+ , and reflected, B_+ , wave components
 123 need to be separated from the total pressure measured at each pressure sensor in the system
 124 shown in Figure 1 because they are used as the reference and error signals in the case of

125 controlling a positive primary source. This separation of wave components has been carried
 126 out using a wave separation method based on the Spatial Fourier Transform (SFT). The
 127 advantages of this approach are that no assumptions are made regarding the nature of the
 128 sound field and it is straightforward to apply to the linear arrays utilised here^{19–22}. In
 129 addition, the SFT has the ability to decompose the pressure distribution of spherical waves
 130 into their plane wave components. For the system shown in Figure 1, a one-dimensional
 131 SFT is required to transform the spatial y -coordinate to the wavenumber domain variable
 132 k_y , which can be written as

$$p(x, k_y) = \int_{-\infty}^{\infty} p(x, y) e^{-jk_y y} dy \quad (1)$$

133 where

$$k_y = k_0 \sin \theta, \quad (2)$$

134 k_y is the component of the wavenumber in the y -direction and θ is the angle of incidence.
 135 Practically, it is not straightforward to apply the SFT given by Eq. 1 due to the integral
 136 of the pressure over an infinite space, however, it has previously been shown that this can
 137 be approximated by the sum of the weighted pressures at several sampling points²². Thus,
 138 the transformed sound pressure at the l -th pressure sensor in the p -th pressure sensor plane,
 139 whose positions are indicated by the variables x_1 to x_4 shown in Figure 1, can be expressed
 140 as²²

$$p_{p_l}(x_{p_l}, y_{p_l}) W(y_{p_l}) \quad (3)$$

141 where

$$W(y_{p_l}) = e^{-jk_y y_{p_l}}, \quad (4)$$

142 subscript p denotes the sensor plane, $W(y_{p_l})$ is the weighting factor that applies the ap-
 143 propriate phase shift corresponding to the angle between the location of the l th pressure
 144 sensor and the plane perpendicular to the incident wave direction, k_0 . When the incident
 145 angle is zero, the pressure sensor planes are already perpendicular to the incident wave di-
 146 rection and, thus, the weighting factors given by Eq. 4 are equal to unity. For time domain
 147 implementation, the ideal impulse responses of the weighting factors can be obtained via
 148 inverse Fourier transform of Eq. 4, however, the impulse responses in this case are non-
 149 casual, which limits their application within a control system. Causality can, however, be
 150 maintained by incorporating modelling delays into the weighting factor responses, which was
 151 previously proposed in²³. This allows the delayed weighting factors to be modelled using
 152 Finite Impulse Response (FIR) filters, which have been designed in this case using a least
 153 mean squares fitting approach²⁴.

154 Using the delayed and weighted pressure measured at each pressure sensor according to
 155 Eq. 3, the same wave separation method described in²⁵ can then be used to calculate the
 156 positive and negative propagating waves at each pair of closely spaced pressure sensors in
 157 the upstream and downstream spaces. In this wave separation method, the total pressure,
 158 p_l , and particle velocity, u_l , at the midpoint of the l -th pair of closely spaced pressure sensors
 159 in the upstream section is calculated as

$$p_l = \frac{p_{1l}W(y_{1l}) + p_{2l}W(y_{2l})}{2}, \quad (5)$$

160 and

$$u_l = \frac{1}{\rho_0 \Delta x} \int_0^{T_s} p_{1l}W(y_{1l}) - p_{2l}W(y_{2l}) dt, \quad (6)$$

161 where p_{1_l} and p_{2_l} are the pressures measured by the sensors at the l -th position in the first
 162 and second sensor planes, T_s is the sampling period and ρ_0 is the density of air. The positive
 163 and negative propagating waves at the l -th pair of closely spaced pressure sensors can be
 164 calculated as

$$A_l = \frac{1}{2} \left(p_l + \rho_0 c_0 u_l \right), \quad (7)$$

165 and

$$B_l = \frac{1}{2} \left(p_l - \rho_0 c_0 u_l \right), \quad (8)$$

166 where c_0 is the speed of sound in air. Substituting Eqs. 5 and 6 into Eqs. 7 and 8, the
 167 positive and negative propagating waves at the l th pair of closely spaced pressure sensors
 168 can also be expressed as

$$\begin{aligned}
 A_l &= \frac{p_{1_l} W(y_{1_l}) + p_{2_l} W(y_{2_l})}{4} \dots \\
 &+ \frac{c_0}{2\Delta x} \int_0^{T_s} p_{1_l} W(y_{1_l}) - p_{2_l} W(y_{2_l}) dt,
 \end{aligned} \quad (9)$$

169 and

$$\begin{aligned}
 B_l &= \frac{p_{1_l} W(y_{1_l}) + p_{2_l} W(y_{2_l})}{4} \dots \\
 &- \frac{c_0}{2\Delta x} \int_0^{T_s} p_{1_l} W(y_{1_l}) - p_{2_l} W(y_{2_l}) dt.
 \end{aligned} \quad (10)$$

170 The wave separation method described in this section has also been applied to the third
 171 and fourth planes of pressure sensors to calculate the wave components, C_l and D_l , at the
 172 l -th pair of pressure sensors in the downstream space and the transmitted wave component
 173 at the l -th pair of closely spaced pressure sensors can be calculated as

$$\begin{aligned}
 C_l &= \frac{p_{3_l} W(y_{3_l}) + p_{4_l} W(y_{4_l})}{4} \dots \\
 &+ \frac{c_0}{2\Delta x} \int_0^{T_s} p_{3_l} W(y_{3_l}) - p_{4_l} W(y_{4_l}) dt,
 \end{aligned} \quad (11)$$

174 where p_{3_l} and p_{4_l} are the pressures measured by the sensors at the l -th position in the third
 175 and fourth sensor planes.

176 IV. WAVE-BASED ACTIVE CONTROL FORMULATION

177 Once the transmitted and reflected wave components have been separated from the total
 178 pressure measured at each pair of pressure sensors using the wave separation method de-
 179 scribed in Section III, these wave components can be used to realise control. The proposed
 180 wave-based active control system uses a multichannel feedforward Filtered-Reference Least
 181 Mean Squares (FxLMS) algorithm to adaptively control the corresponding wave components
 182 and the block diagram for this active control system is shown in Figure 2. The FxLMS al-
 183 gorithm is the most widely utilised adaptive feedforward control strategy for active noise
 184 control and its implementation and operation has been widely discussed²⁶. The vector of $2L$
 185 error signals is generated by combining the L reflected, B_{+l} , and L transmitted, C_{+l} , wave
 186 components, whilst the positive propagating incident wave, A_{+l} , for the L pairs of pressure
 187 sensors are used to generate the vector of L reference signals. As mentioned in Section I, the
 188 advantage of the proposed wave-based actively controlled metasurface is that the direction
 189 of the nonreciprocal behaviour can be straightforwardly reversed by changing the reference
 190 and error signals used by the wave-based controller shown in Figure 2, however, this is not
 191 demonstrated explicitly here for conciseness.

192 In the considered configuration, the dual layer array of control sources are driven to
 193 minimise the transmitted, C_{+l} , and reflected, B_{+l} , wave components with respect to the
 194 incident acoustic field generated by the positive primary source, whilst allowing the incident

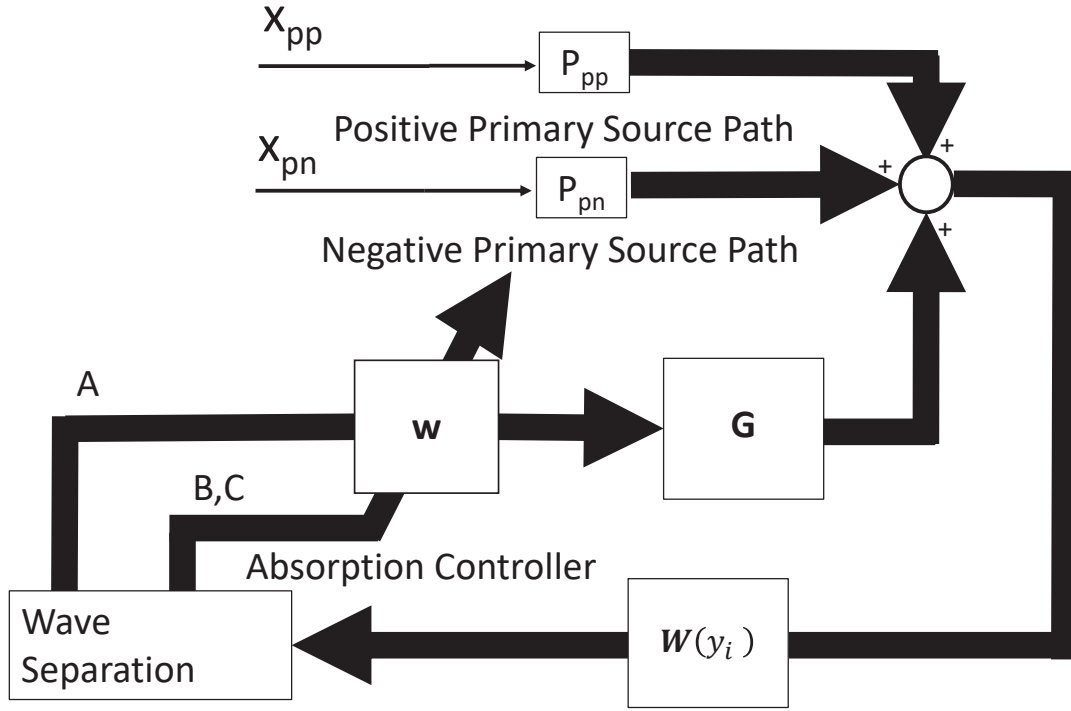


FIG. 2. The block diagram of the wave-based active controller used to drive the dual-layer array of control sources forming the metasurface to achieve nonreciprocal sound absorption.

195 acoustic field generated by the negative primary source to propagate freely, thus achieving
 196 nonreciprocal sound absorption. The vector of error signals in this case can be defined at
 197 the n -th time sample as

$$\mathbf{e}(n) = \mathbf{d}(n) + \mathbf{R}(n)\mathbf{w}(n) \quad (12)$$

198 where

$$\mathbf{e}(n) = \begin{bmatrix} e_{T_1}(n) & e_{R_1}(n) & \dots & e_{T_L}(n) & e_{R_L}(n) \end{bmatrix}^T, \quad (13)$$

199 is the vector of $2L$ error signals;

$$\mathbf{d}(n) = \begin{bmatrix} d_{T_1}(n) & d_{R_1}(n) & \dots & d_{T_L}(n) & d_{R_L}(n) \end{bmatrix}^T, \quad (14)$$

200 is the vector of $2L$ disturbance signals comprised of the transmitted and reflected wave com-

201 ponents generated by the positive primary source, where the l -th reflected and transmitted
 202 wave components are given by Eqs. 10 and 11 respectively;

$$\mathbf{w}(n) = \left[w_{11_i}(n), w_{12_i}(n), \dots, w_{MKI}(n) \right]^T \quad (15)$$

203 is the $2MKI$ vector of FIR control filter coefficients, where M is the number of control
 204 sources, K is the number of reference signals, which is equal to L in this case and I is the
 205 length of each FIR control filter;

$$\mathbf{R}(n) = \begin{bmatrix} \mathbf{r}_{T_1}(n) & \mathbf{r}_{T_1}(n-1) & \dots & \mathbf{r}_{T_1}(n-I-1) \\ \mathbf{r}_{R_1}(n) & \mathbf{r}_{R_1}(n-1) & \dots & \mathbf{r}_{R_1}(n-I-1) \\ \mathbf{r}_{T_2}(n) & \mathbf{r}_{T_2}(n-1) & \dots & \mathbf{r}_{T_2}(n-I-1) \\ \mathbf{r}_{R_2}(n) & \mathbf{r}_{R_2}(n-1) & \dots & \mathbf{r}_{R_2}(n-I-1) \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{r}_{T_L}(n) & \mathbf{r}_{T_L}(n-1) & \dots & \mathbf{r}_{T_L}(n-I-1) \\ \mathbf{r}_{R_L}(n) & \mathbf{r}_{R_L}(n-1) & \dots & \mathbf{r}_{R_L}(n-I-1) \end{bmatrix}, \quad (16)$$

206 is the $(2L \times MKI)$ matrix of transmitted and reflected wave components calculated from the
 207 filtered reference signals, where the l -th vector of reflected, $\mathbf{r}_{R_l}(n)$, and transmitted, $\mathbf{r}_{T_l}(n)$,
 208 wave components have been defined as

$$\mathbf{r}_{R_l}(n) = \left[r_{R_{l1}}(n) \ r_{R_{l2}}(n) \ \dots \ r_{R_{lMK}}(n) \right]^T, \quad (17)$$

209 and

$$\mathbf{r}_{T_l}(n) = \left[r_{T_{l1}}(n) \ r_{T_{l2}}(n) \ \dots \ r_{T_{lMK}}(n) \right]^T, \quad (18)$$

210 where $r_{R_{lmk}}(n)$ and $r_{T_{lmk}}(n)$ are the reflected and transmitted wave components calculated
 211 from the filtered reference signals according to Eqs. 10 and 11 respectively. The reference
 212 signals have been filtered by the FIR filter that represents the plant response between the
 213 m -th control source and l -th pressure sensor within the p -th sensor plane, which have been
 214 calculated based on an initial identification phase as standard in FxLMS implementations.
 215 These filtered reference signals are given by

$$r_{plm} = \sum_{j=0}^{J-1} g_{plmj} A_{+l}(n-j), \quad (19)$$

216 where A_{+l} is the l -th reference signal corresponding to the incident wave component and
 217 g_{plmj} is the j -th FIR filter coefficient of the J coefficient filter representing the plant response
 218 between the m -th control source and the l -th pressure sensor within the p -th plane of pressure
 219 sensors. The cost function in this case is the sum of the mean squared error signals, which
 220 can be expressed as

$$J(n) = \mathbb{E} \left[\mathbf{e}^T(n) \mathbf{e}(n) \right]. \quad (20)$$

221 Substituting Eq. 12 into Eq. 20, the cost function can be expressed in Hermitian quadratic
 222 form as

$$\begin{aligned}
 J(n) &= \mathbf{w}^T(n) \mathbf{R}^T(n) \mathbf{R}(n) \mathbf{w}(n) \dots \\
 &+ 2\mathbf{w}^T(n) \mathbf{R}^T(n) \mathbf{d}(n) + \mathbf{d}^T(n) \mathbf{d}(n).
 \end{aligned} \quad (21)$$

223 Taking the derivative of the cost function with respect to the vector of FIR control filter
 224 coefficients, the resulting gradient can be expressed as

$$\begin{aligned}
 \frac{\partial J(n)}{\partial \mathbf{w}(n)} &= 2[\mathbf{R}^T(n) \mathbf{R}(n) \mathbf{w}(n) + \mathbf{R}^T(n) \mathbf{d}(n)] \dots \\
 &= 2\mathbf{R}(n) \mathbf{e}(n).
 \end{aligned} \quad (22)$$

225 Using the negative gradient given by Eq. 22, the multichannel FxLMS algorithm can be
 226 used to adapt the vector of FIR control filter coefficients to minimise the transmitted and
 227 reflected wave components as

$$\mathbf{w}(n+1) = \mathbf{w}(n) - \mu \mathbf{R}(n) \mathbf{e}(n), \quad (23)$$

228 where μ is the convergence gain, which controls the speed and stability of the adaptation.

229 V. SIMULATION STUDY

230 In order to investigate the performance of the proposed actively controlled metasurface
 231 under ideal conditions, this section presents the results from a simulation based investiga-
 232 tion, where the control sources and primary sources are modelled as monopoles in a free field
 233 environment and the sensors are assumed to be omnidirectional. The performance of the
 234 proposed wave-based active controller described in Section IV is evaluated when it is subject
 235 to an incident sound field that is generated by either normally or obliquely-positioned pos-
 236 itive or negative primary sources. The performance of the proposed controller is evaluated
 237 using the magnitude of the pressure transmission and reflection coefficients averaged across
 238 the array of L , which is defined for a positive incident primary source as

$$T = \frac{1}{L} \sum_{l=1}^L \left| \frac{C_{+l}}{A_{+l}} \right|, \quad (24)$$

$$R = \frac{1}{L} \sum_{l=1}^L \left| \frac{B_{+l}}{A_{+l}} \right| \quad (25)$$

239 and the power absorption coefficient defined using the averaged pressure transmission and
 240 reflection coefficients as

$$\alpha = 1 - (|T|^2 + |R|^2). \quad (26)$$

241 Despite the fact that the system shown in Figure 1 only considers a two-dimensional plane,
 242 the incident sound field produced by both primary sources has been assumed to radiate
 243 spherically into three dimensional space, in order to be consistent with the experimental
 244 configuration described in Section II. As a result, the incident sound field and performance
 245 metrics in the uncontrolled case encounter losses due to spherical spreading. To ensure
 246 consistency between the theoretical and experimental results, the theoretically-modelled
 247 acoustic responses between each acoustic source and each pressure sensor have been modelled
 248 as outgoing spherical waves, which can be expressed as

$$p(r) = \frac{p_0}{|r|} e^{-jk_0|r|} \quad (27)$$

249 where r is the distance between the acoustic source and the pressure sensor, p_0 is the pressure
 250 amplitude and $|\cdot|$ is an Euclidean norm. The losses due to spherical spreading are included in
 251 these simulated acoustic responses via the $\frac{1}{|r|}$ term in Eq. 27. To demonstrate nonreciprocal
 252 behaviour, the performance metrics in the controlled and uncontrolled cases are compared
 253 for both positive and negative incident primary fields, in the case of normal or obliquely
 254 positioned primary sources, and these results are presented in the following subsections.

255 **A. Normally-Positioned Primary Source**

256 In the first instance, the performance metrics have been calculated before and after im-
 257 plementing the proposed wave-based active controller when the fields incident on the meta-
 258 surface are generated by the normally-positioned positive and negative primary sources and

259 these results are presented in Figure 3. The solid and dashed lines in Figure 3 show the
260 magnitude of the average transmission (blue lines), reflection (red lines) and absorption
261 (black lines) coefficients for the controlled and uncontrolled cases respectively. Figure 3(a)
262 shows the behaviour for the positive primary source and Figure 3(b) shows the behaviour
263 for the negative primary source. As noted in Section IV, the proposed active metasurface
264 has been configured to control the wave components with respect to the positive primary
265 source, whilst allowing the incident sound field generated by the negative primary source
266 to propagate unimpeded. It is worth initially highlighting that the performance metrics in
267 the uncontrolled cases for both positive and negative normally-positioned primary sources
268 are identical, demonstrating the conventional reciprocal behaviour prior to control. It is
269 also worth reiterating that the absorption coefficient is non-zero in the uncontrolled case
270 due to the spherical spreading that occurs in the three-dimensional simulated environment.
271 With control, it can be seen from the presented results that the absorption controller achieves
272 near-zero transmission and reflection coefficients with respect to the positive primary source,
273 which leads to near-perfect sound absorption, whilst the performance metrics in the con-
274 trolled case are equal to the uncontrolled metrics with respect to the negative primary
275 source as shown in Figure 3(b). These results show that the active metasurface achieves
276 nonreciprocal sound absorption when subject to normally-positioned primary sources.

277 To provide further insight into how the metasurface influences the total sound pressure
278 field, the pressure contour plots in the uncontrolled and controlled cases at 400 Hz are
279 presented in Figure 4. Figure 4(a) shows the uncontrolled incident sound field generated
280 by the normally-positioned positive primary source and Figure 4(b) shows the controlled

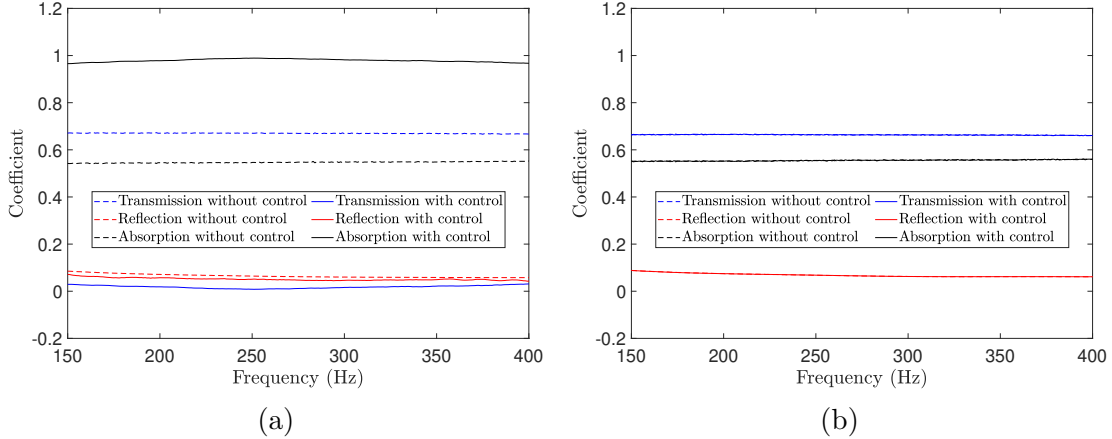


FIG. 3. The performance of the active metasurface with (solid lines) and without (dashed lines) control in terms of the transmission (blue lines), reflection (red lines) and absorption (black lines) coefficients when the incident waves are produced by the positive (a) and negative (b) primary sources located at the normal.

281 pressure field. These results show the active metasurface minimises both the transmitted
 282 and reflected wave components, leading to near-perfect sound absorption, but the control is
 283 clearly limited to within the aperture of the sensor and control source arrays.

284 B. Obliquely-Positioned Primary Source

285 In the two-dimensional case considered here, it is also important to investigate the
 286 performance of the wave-based actively controlled metasurface when subject to obliquely-
 287 positioned positive and negative primary sources. In the case of obliquely-positioned primary
 288 sources, the sources have a 45° and -45° angle of incidence respectively, as shown in Fig-
 289 ure 1. Similarly to Section V A, the performance metrics for the controlled (solid lines) and
 290 uncontrolled (dashed lines) cases have been calculated and these results are presented in Fig-
 291 ure 5. Figure 5(a) shows that minimising the transmitted and reflected wave components

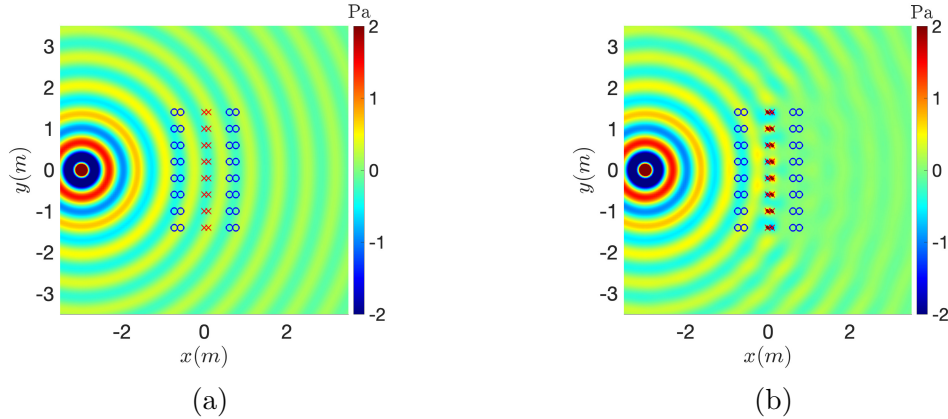


FIG. 4. The pressure contour plots of the incident sound field corresponding to the pressure generated by the normally-positioned positive primary source without control (a) and with control (b) at 400Hz. The red crosses indicate the monopole control source positions and the blue circles indicate the pressure sensor positions.

292 maximises the absorption coefficient and it is close to 0.9 across the presented bandwidth.
 293 In contrast to the positive primary source case shown in Figure 5(a), the behaviour of the
 294 absorption controller with respect to the obliquely-positioned negative primary source shows
 295 that the controlled metrics are similar to the uncontrolled case, with some small differences
 296 at lower and higher frequencies within the presented bandwidth. These results show that the
 297 wave attenuation achieved by the active metasurface in the oblique source case is reduced
 298 compared to the normally-positioned case, however, the actively controlled metasurface still
 299 achieves significant levels of nonreciprocal sound absorption for the obliquely-positioned
 300 primary sources.

301 As in the case of the normally-positioned primary source, it is insightful to investigate
 302 how the active metasurface influences the sound pressure field in more detail and the pres-
 303 sure contour plots are presented in Figure 6. Figure 6(a) shows the incident sound field

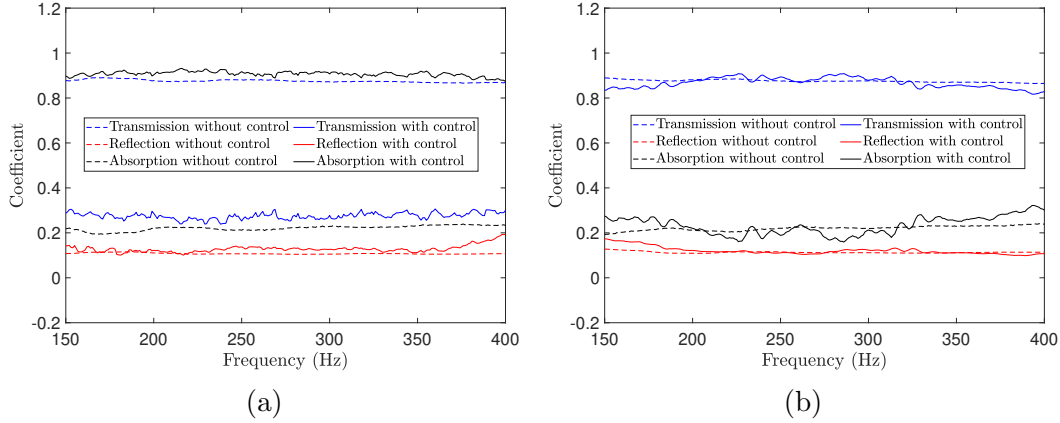


FIG. 5. The performance of the active metasurface with (solid lines) and without (dashed lines) control in terms of the transmission (blue lines), reflection (red lines) and absorption (black lines) coefficients when the spherical incident waves are produced by the positive (a) and negative (b) primary sources obliquely-positioned.

304 generated by the obliquely-positioned positive primary source and Figure 6(b) shows the
 305 controlled sound field. From these plots it can be seen that the actively controlled metasur-
 306 face minimises both the transmission and reflection, which leads to the incident wave being
 307 absorbed. In the case of the obliquely-positioned primary source, however, it can be seen
 308 that diffraction occurs at the edge of the control source array closer to the primary source.

309 VI. EXPERIMENTAL VALIDATION

310 The simulation results presented in Section V show that the actively controlled metasur-
 311 face has the ability to control the various wave components to achieve nonreciprocal sound
 312 absorption in a two-dimensional plane. It is important to experimentally validate the simu-
 313 lation results to provide insight into the effects of imperfect free-field conditions, finite-sized
 314 sensors and finite-sized loudspeakers, which are used to realise the control sources in a prac-

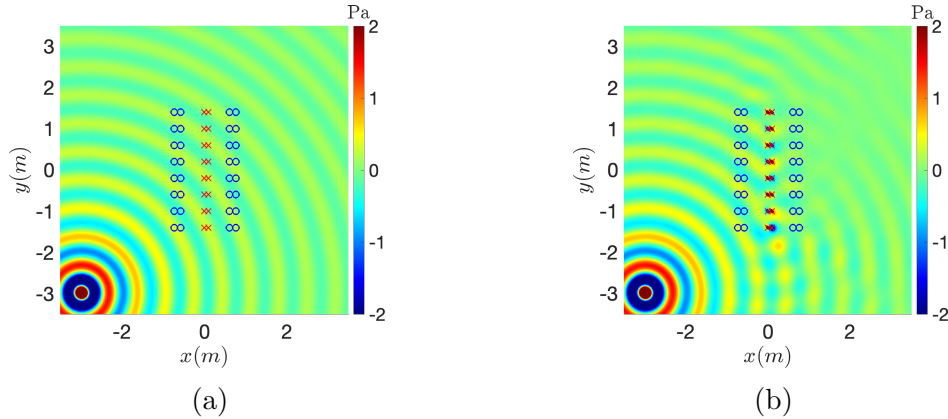


FIG. 6. The pressure contour plot of the incident sound field corresponding to the pressure generated by the obliquely-positioned primary source without control (a) and with control (b) at 400Hz. The red crosses indicate the monopole control source positions and the blue circles indicate the pressure sensor positions.

315 tical system. As mentioned in Section I, the experimental validation has been carried out
 316 through offline time-domain simulations using measured responses obtained from a practical
 317 system constructed in the anechoic chamber at the ISVR and this system will be described
 318 in the following section.

319 A. Experimental System Description

320 Figure 7 shows the practical system that has been used to experimentally validate the
 321 simulation results presented in Section V. The system shown in Figure 7 has the same
 322 physical arrangement and number of sensors and sources as the simulated system shown in
 323 Figure 1, with primary sources positioned at normal and oblique angles to the control and
 324 sensor arrays, a dual-layer array of control sources located at the centre and two dual-layer
 325 arrays of pressure sensors positioned either side of the control sources. The pairs of control

326 sources have been practically realised using two sealed-back Visaton B80 loudspeakers, which
327 have a cone diameter of 8 cm, and share the same enclosure as shown in Figure 7(c). The
328 primary sources have been realised using JBL Control 1 Pro full-range loudspeakers and
329 the microphone arrays have been implemented using 1/4-inch array microphones. All of
330 the data acquisition has been carried out using a Dante-enabled system, which drives each
331 acoustic source individually, via a reconstruction filter and an amplifier, with a logarithmic
332 sine sweep to obtain the impulse responses between each acoustic source and each pressure
333 sensor. The logarithmic sine sweep method has been used because it achieves a higher signal
334 to noise ratio in an anechoic environment and is less time consuming than using white noise
335 to measure the large number of impulse responses²⁸. A Larson Davis Cal250 Sound Level
336 Calibrator has been used to calibrate all of the pressure sensors, which are connected via
337 signal conditioning and antialiasing filters to the Analogue to Digital Converters.

338 B. Results

339 Having acquired the responses between the primary sources, control sources and arrays
340 of pressure sensors, the performance of the proposed active metasurface and absorption con-
341 trol strategy has been evaluated using offline simulations with the measured responses; this
342 means that the real-time processing requirements have not been considered here, but the
343 physical acoustic effects of the practical implementation have been taken into account. The
344 active metasurface performance has been evaluated, as in the case of the theoretical simula-
345 tions, for both normally and obliquely-positioned positive and negative primary sources and
346 the results are presented in Figures 8 and 9 respectively. As in Section V, the performance

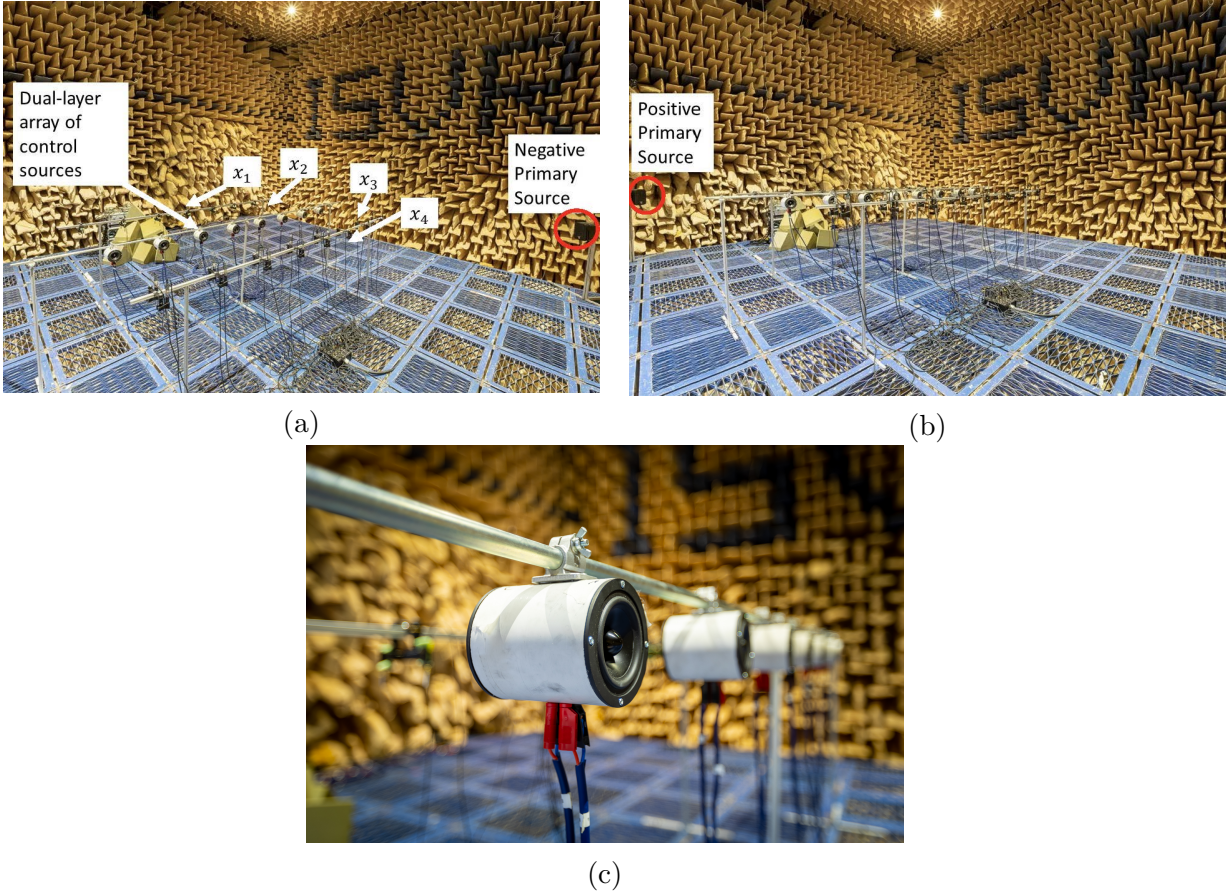


FIG. 7. (a) and (b) show the practical realisation used to obtain the measured response data required to experimentally validate the proposed active metasurface. (c) shows one of the sub-wavelength active unit cells consisting of a pair of sealed back loudspeakers.

347 metrics before (dashed lines) and after (solid lines) implementing the proposed wave-based
 348 active controller have been calculated in each considered case. The experimental results
 349 presented in Figures 8 and 9 are largely consistent with the simulation results presented in
 350 Section V, confirming the ability of the proposed active metasurface to achieve nonreciprocal
 351 sound absorption in a two-dimensional plane. For both the normal and obliquely positioned
 352 configurations, the experimental results show greater fluctuations over frequency in the
 353 performance metrics compared to the theoretical simulations, which can be related to the

354 presence of the finite-sized sources and sensors introducing scattering into the environment.
 355 However, in both cases it is still clear that the positive incident wave is largely absorbed,
 356 while the negative incident wave is allowed to pass without significant modification.

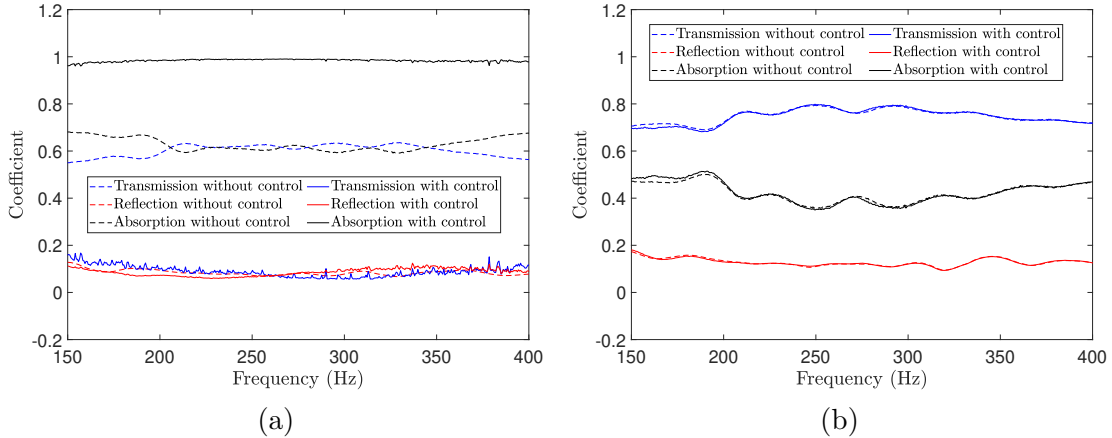


FIG. 8. The performance of the experimental active metasurface with (solid lines) and without (dashed lines) control in terms of the transmission (blue lines), reflection (red lines) and absorption (black lines) coefficients when the incident sound field is generated by the normally-positioned positive (a) and negative (b) primary sources.

357 VII. CONCLUSIONS

358 The work presented in this paper has demonstrated through theoretically-modelled and
 359 offline experimental simulation-based investigations that the feedforward wave-based active
 360 controller and single subwavelength unit cell previously proposed and explored for one-
 361 dimensional environments¹⁶ can be extended to a metasurface consisting of a dual-layer
 362 array of control sources to control the transmitted and reflected wave components to achieve
 363 nonreciprocal sound absorption in a two-dimensional plane within a three-dimensional space.

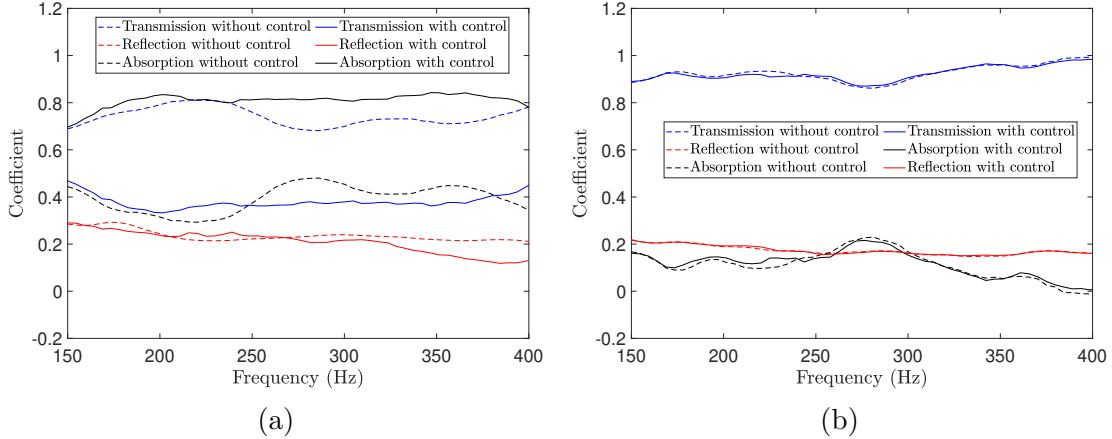


FIG. 9. The performance of the experimental active metasurface with (solid lines) and without (dashed lines) control in terms of the transmission (blue lines), reflection (red lines) and absorption (black lines) coefficients when the incident sound field is generated by the obliquely-positioned positive (a) and negative (b) primary sources.

364 A wave separation method, which is based on the discrete Spatial Fourier Transform, has
 365 been proposed to separate the positive and negative propagating waves in the upstream
 366 and downstream of the active metasurface consisting of the array of control sources and
 367 arrays of error and reference sensors. A wave-based control strategy for the two-dimensional
 368 control problem has then been outlined, which utilises multiple control sources to realise
 369 nonreciprocal control over a region of space.

370 The performance of the wave-based actively controlled metasurface has been evaluated
 371 when subject to incident acoustic fields that are generated either by normally or obliquely-
 372 positioned positive and negative primary sources. The performance metrics in the controlled
 373 and uncontrolled cases have been compared to evaluate the nonreciprocal behaviour of the
 374 metasurface. Although the performance of the proposed metasurface is slightly less in the

375 oblique case compared to the normal case, it has been shown that the proposed approach
376 still achieves nonreciprocal sound absorption over the presented bandwidth.

377 One advantage of the proposed actively controlled metasurface is that it is reconfigurable,
378 such that the direction of the nonreciprocal behaviour can be reversed by changing the ref-
379 erence and error signals used by the controller, which simply requires changing the signals
380 in the control system rather than any physical modifications. In addition to the reconfigura-
381 bility of the nonreciprocal control approach explored in this paper, it is important for future
382 work to explore how robust the proposed wave-based control strategy is to deviations in the
383 exact positioning of both the sources and sensors forming the active metasurface. Moreover,
384 although in certain situations it would be possible to control multiple incident waves using
385 multiple iterations of the control strategy proposed here, this would not be straightforwardly
386 realisable for the independent control of multiple coherent sources. Therefore, further work
387 is also required to explore the challenges associated with independently controlling multiple
388 incident sources under all circumstances.

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393 **AUTHOR DECLARATIONS**

394 Jordan Cheer and Joe Tan have patent WO2023089300A1 and WO2023089301A1 pend-
395 ing.

396 **DATA AVAILABILTY**

397 The data that support the findings of this study are available from the corresponding
398 author upon reasonable request.

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